

# Imaging an acoustic waveguide from surface data in the time domain

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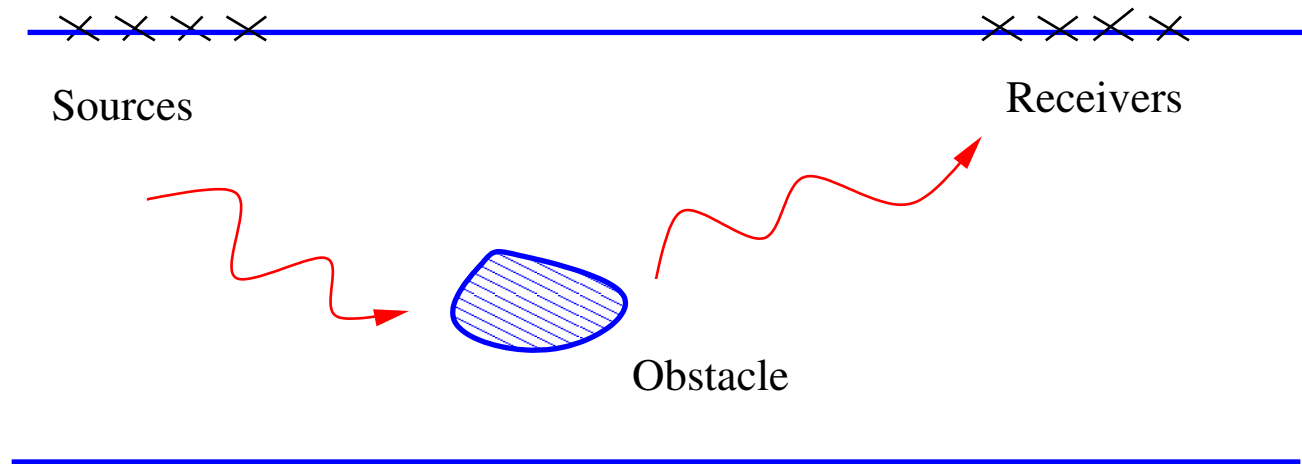
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# Statement of the inverse problem

**Geometry** : a 2D acoustic waveguide in the time domain



**Objective** : find a sound soft obstacle from measurements at surface receivers of scattered waves due to surface sources

# A brief state of the art

## Sampling Methods in waveguides:

- **Frequency domain**

**Acoustics** : Xu, Mawata & Lin (2000) : LSM, Charalambopoulos, Gintides, Kiriaki & Kirsch (2006), Arens, Gintides & Lechleiter (2011) : Factorization Method, Bourgeois & Lunéville (2008) : Modal approach, Monk & Selgas (2012) : RG-LSM

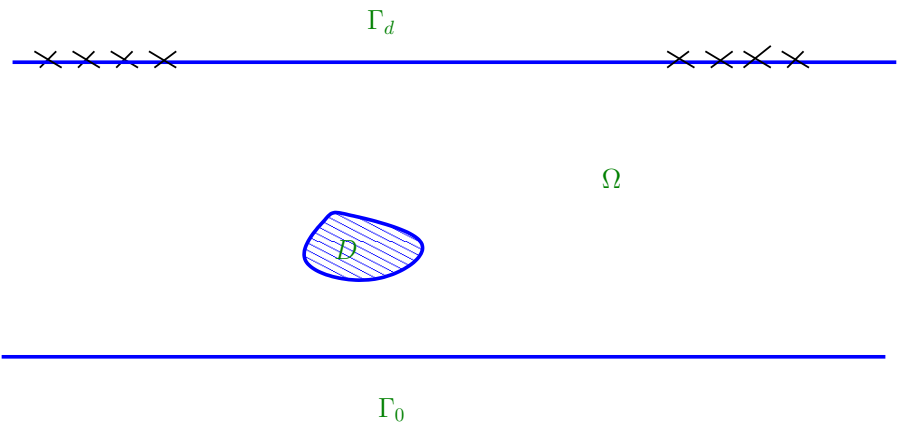
**Elasticity** : Bourgeois, Le Louër & Lunéville (2011)

**Electromagnetism** : Borcea & Nguyen (2016)

- **Time domain**

Monk & Selgas (2016) : first tentative

# The forward problem in the time domain



$$\left\{ \begin{array}{lll} \frac{1}{c^2} \partial_t^2 v - \Delta v & = & 0 \quad \text{in } \Omega \times (0, +\infty) \\ \partial_\nu v & = & f \chi \quad \text{on } \Gamma_d \times (0, +\infty) \\ \partial_\nu v & = & 0 \quad \text{on } \Gamma_0 \times (0, +\infty) \\ v & = & 0 \quad \text{on } \partial D \times (0, +\infty) \\ v, \partial_t v & = & 0 \quad \text{on } \Omega \times \{0\} \end{array} \right.$$

**Data :** space function  $f = \delta(\cdot - x_1^{m\pm})$  with  $x_1^{m\pm} = \pm(R + m\delta)$  and  $m = 0, \dots, M - 1$ , time function  $\chi$  compactly supported in  $[0, +\infty)$   
 $\longrightarrow$  correctly choose  $M$ ,  $\delta$  and  $\chi$  !

# The forward problem in the frequency domain

Apply the **Fourier transform** :

$$u := \widehat{v}(x, \omega) = \int_{\mathbb{R}} v(x, t) e^{i\omega t} dt$$

$$\left\{ \begin{array}{ll} (\Delta + k^2)u = 0 & \text{in } \Omega \\ \partial_\nu u = \phi & \text{on } \Gamma_d \\ \partial_\nu u = 0 & \text{on } \Gamma_0 \\ u = 0 & \text{on } \partial D \\ + \text{Radiation condition} \end{array} \right.$$

with

$$k = \omega/c \quad \phi = \widehat{\chi}(\omega)f$$

A problem in the time domain  $\longrightarrow$  Harmonic problems for multiple frequencies

# The guided modes

- The guided modes : find  $u$  such that

$$\begin{cases} (\Delta + k^2)u = 0 & \text{in } W \\ \partial_\nu u = 0 & \text{on } \Gamma \end{cases}$$

- $\theta_n$  and  $\lambda_n$  ( $n \in \mathbb{N}$ ) : Neumann eigenfunctions and eigenvalues of the 1D operator  $-\Delta$  in transverse section  $\Sigma$
- The  $\theta_n$  form a complete basis of  $L^2(\Sigma)$  while  $\lambda_0 = 0 < \lambda_1 < \lambda_2 < \lambda_3 < \dots < \lambda_n \longrightarrow +\infty$
- Guided modes :  $u_n^\pm(x_1, x_2) = \theta_n(x_2)e^{\pm i\beta_n x_1}$ ,  $\beta_n = \sqrt{k^2 - \lambda_n}$  for  $n = 0, \dots, N-1$  (propagating modes) and  $\beta_n = i\sqrt{\lambda_n - k^2}$  for  $n \geq N$
- Assumption on  $k$  :  $\beta_n \neq 0$  for all  $n \in \mathbb{N}$

# The fundamental solution

For  $y \in W$ , the solution to

$$\left\{ \begin{array}{ll} -(\Delta + k^2)G(\cdot, y) = \delta_y & \text{in } W \\ \partial_\nu G(\cdot, y) = 0 & \text{on } \Gamma_0 \cup \Gamma_d \\ + \text{Radiation condition} \end{array} \right.$$

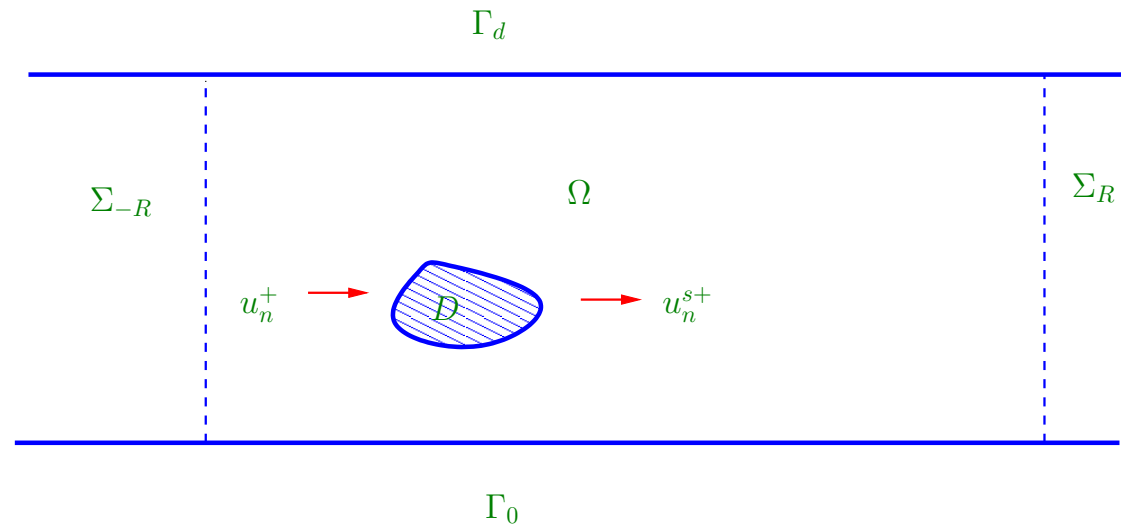
and for  $y \in \Gamma_d$ , the solution to

$$\left\{ \begin{array}{ll} -(\Delta + k^2)G(\cdot, y) = 0 & \text{in } W \\ \partial_\nu G(\cdot, y) = \delta_y & \text{on } \Gamma_d \\ \partial_\nu G(\cdot, y) = 0 & \text{on } \Gamma_0 \\ + \text{Radiation condition} \end{array} \right.$$

are given by

$$G(x, y) = - \sum_{n \in \mathbb{N}} \frac{e^{i\beta_n |x_1 - y_1|}}{2i\beta_n} \theta_n(x_2) \theta_n(y_2)$$

# An ideal inverse problem



**Inverse problem with guided modes** : we measure on  $\hat{\Sigma} = \Sigma_{-R} \cup \Sigma_{+R}$  the scattered fields  $u_n^{s\pm}$  associated to the incident fields  $u_n^\pm$  for all  $n \in \mathbb{N}$  : find  $D$

**Projection of scattered fields on  $\hat{\Sigma}$**  :

$$u_n^{s+}|_{\Sigma_{-R}} = \sum_{m \in \mathbb{N}} S_{mn}^{+-} \theta_m, \quad u_n^{s+}|_{\Sigma_{+R}} = \sum_{m \in \mathbb{N}} S_{mn}^{++} \theta_m,$$

$$u_n^{s-}|_{\Sigma_{-R}} = \sum_{m \in \mathbb{N}} S_{mn}^{--} \theta_m, \quad u_n^{s-}|_{\Sigma_{+R}} = \sum_{m \in \mathbb{N}} S_{mn}^{-+} \theta_m$$



# The Linear Sampling Method

**Theorem :** for  $z \in W$ , if  $G(\cdot, z)|_{\hat{\Sigma}} \in R(\mathcal{N})$  then  $z \in D$

Equation  $\mathcal{N}h = G(\cdot, z)|_{\hat{\Sigma}}$  in  $L^2(\hat{\Sigma})$  is equivalent to

$$\forall m \in \mathbb{N}, \quad \begin{cases} \sum_{n \in \mathbb{N}} \frac{e^{i\beta_n R}}{2i\beta_n} \left( S_{mn}^{+-} h_n^- + S_{mn}^{--} h_n^+ \right) = \frac{e^{i\beta_m(R+z_1)}}{2i\beta_m} \theta_m(z_2) \\ \sum_{n \in \mathbb{N}} \frac{e^{i\beta_n R}}{2i\beta_n} \left( S_{mn}^{++} h_n^- + S_{mn}^{-+} h_n^+ \right) = \frac{e^{i\beta_m(R-z_1)}}{2i\beta_m} \theta_m(z_2), \end{cases}$$

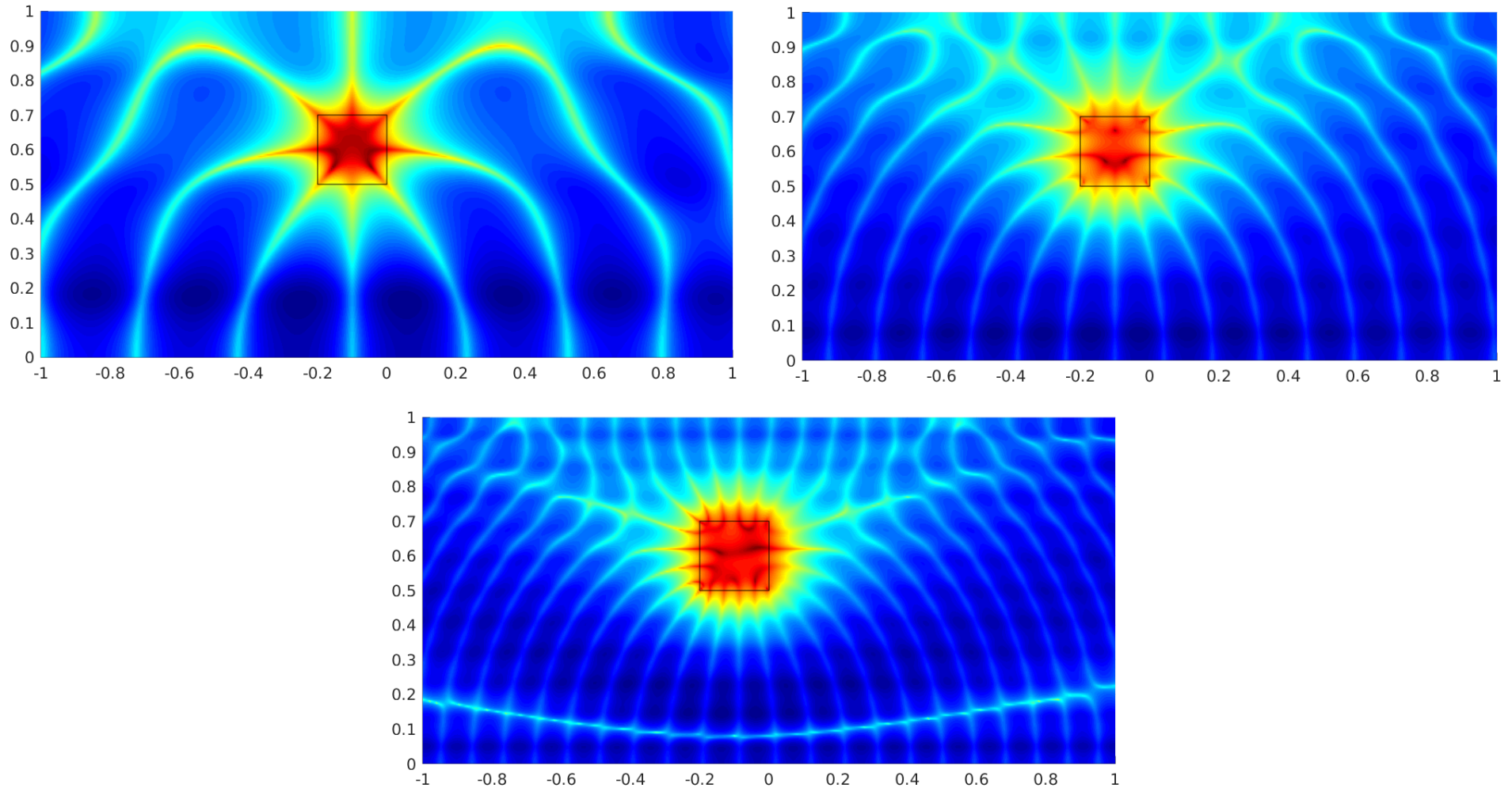
with

$$z = (z_1, z_2), \quad h = (h^-, h^+), \quad h^- = \sum_{n \in \mathbb{N}} h_n^- \theta_n, \quad h^+ = \sum_{n \in \mathbb{N}} h_n^+ \theta_n$$

**Indicator function :**  $\psi(z) = 1/\|h(z)\|_{L^2(\hat{\Sigma})}$  characterizes the defect

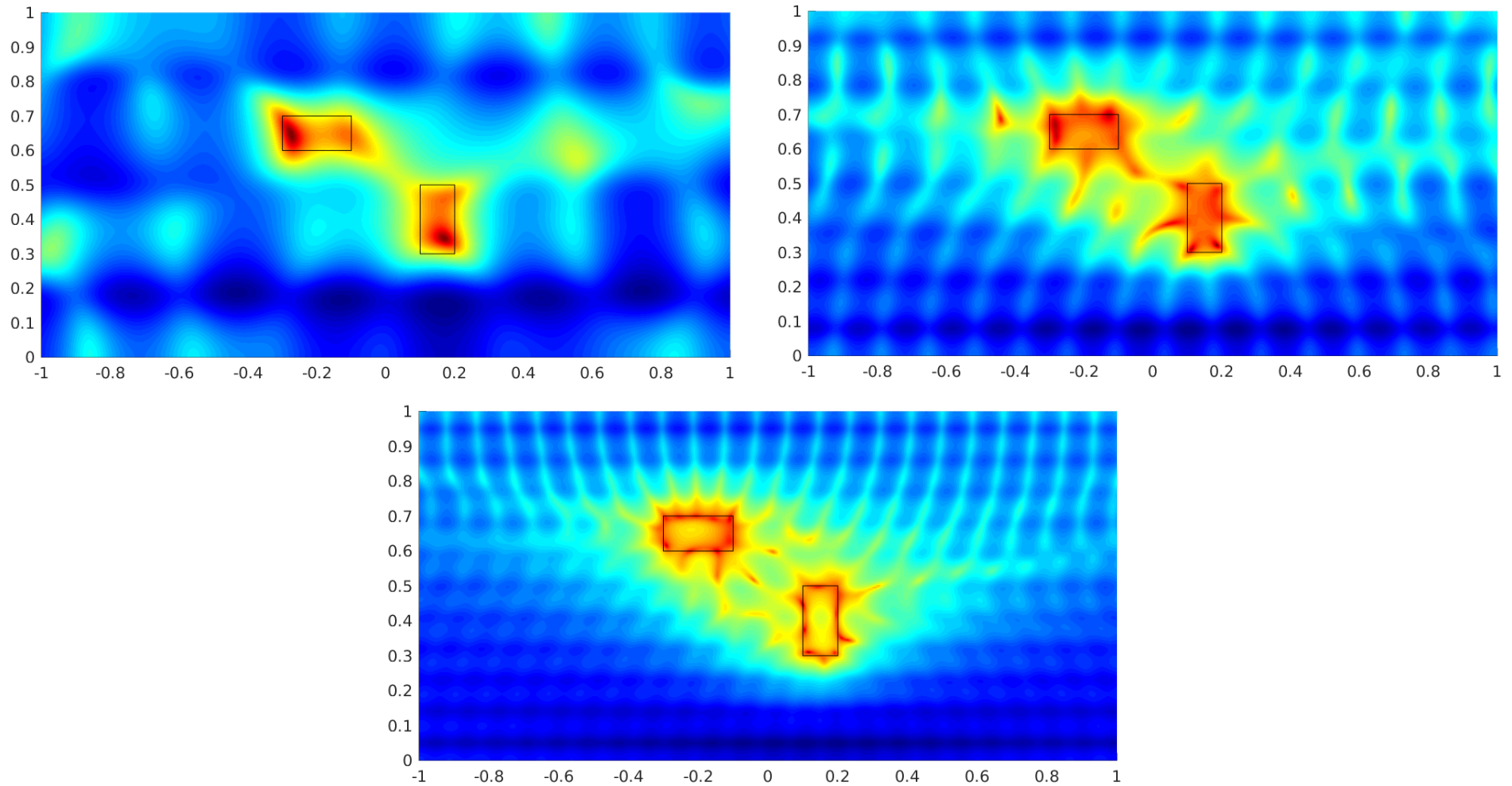
**Discretization :**  $m, n = 0, \dots, N-1$  ( $N$  : number of propagating modes)  $\longrightarrow$  scattering matrix  $S$ , LSM matrix  $U_{mn} = (e^{i\beta_n R}/2i\beta_n)S_{mn}$

# Some numerical results : frequency domain



Number of propagating modes :  $N = 4$ ,  $N = 8$  and  $N = 12$

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# The case of surface data

$$\left\{ \begin{array}{ll} (\Delta + k^2)u = 0 & \text{in } \Omega \\ \partial_\nu u = \phi & \text{on } \Gamma_d \\ \partial_\nu u = 0 & \text{on } \Gamma_0 \\ u = 0 & \text{on } \partial D \\ + \text{ Radiation condition} \end{array} \right.$$

Incident field for  $\phi = \delta(\cdot - y_1)$ , left source and right receiver :

$$\begin{aligned} u^i(x_1, x_2) &= - \sum_{n \in \mathbb{N}} \frac{e^{i\beta_n |x_1 - y_1|}}{2i\beta_n} \theta_n(x_2) \theta_n(d) \\ &= - \sum_{n \in \mathbb{N}} \frac{e^{-i\beta_n y_1}}{2i\beta_n} \theta_n(d) \underbrace{e^{i\beta_n x_1} \theta_n(x_2)} \\ &= - \sum_{n \in \mathbb{N}} \frac{e^{-i\beta_n y_1}}{2i\beta_n} \theta_n(d) u_n^+(x_1, x_2) \end{aligned}$$

# The case of surface data (cont.)

Scattered field  $u^s = u - u^i$  :

$$u^s(x_1, x_2) = - \sum_{n \in \mathbb{N}} \frac{e^{-i\beta_n y_1}}{2i\beta_n} \theta_n(d) u_n^{s+}(x_1, x_2)$$

For  $y_1 = -R - s\delta$ ,

$$u^s(x_1, x_2) = - \sum_{n \in \mathbb{N}} \frac{e^{i\beta_n R}}{2i\beta_n} \theta_n(d) e^{is\beta_n \delta} u_n^{s+}(x_1, x_2)$$

For  $x_1 = R + r\delta$ ,

$$u_n^{s+}(x_1, d) = \sum_{m \in \mathbb{N}} (u_n^{s+}(R, \cdot), \theta_m)_R e^{i\beta_m(x_1 - R)} \theta_m(d) = \sum_{m \in \mathbb{N}} S_{mn}^{++} e^{ir\beta_m \delta} \theta_m(d)$$

Then

$$u^s(x_1, d) = - \sum_{m \in \mathbb{N}} \sum_{n \in \mathbb{N}} e^{ir\beta_m \delta} \theta_m(d) \underbrace{\frac{e^{i\beta_n R}}{2i\beta_n} S_{mn}^{++} \theta_n(d) e^{is\beta_n \delta}}_{U_{mn}^{++}}$$

# The measurement matrix

The measurement matrix : for  $r, s = 0, \dots, M - 1$ ,

$$M_{rs}^{++} = - \sum_{m=0}^{N-1} \sum_{n=0}^{N-1} e^{ir\beta_m\delta} \theta_m(d) U_{mn}^{++} \theta_n(d) e^{is\beta_n\delta}$$

Factorization

$$M = -RUE^t$$

- $M$  : measurement matrix
- $R, E$ : reception and emission matrices
- $U$  : LSM matrix

For  $\phi = \delta(\cdot - y_1)$ ,  $R = E = VT$  with  $T = \text{Diag}(\theta_n(d))$  and  $V$  is the Vandermonde Matrix

$$V_{mn} = e^{im\beta_n\delta}, \quad m = 0, \dots, M - 1, \quad n = 0, \dots, N - 1$$

# Optimization of sources/receivers

**We have to invert  $V^*V$  for  $M \geq N$  :** how to choose  $M$  and  $\delta$  ?

Vandermonde Matrix  $V = (V_{mn})$ ,  $m = 0, \dots, M-1$ ,  $n = 0, \dots, N-1$  :

$$V = \begin{bmatrix} 1 & 1 & \dots & 1 \\ e^{2\pi i f_0} & e^{2\pi i f_1} & \dots & e^{2\pi i f_{N-1}} \\ \dots & \dots & \dots & \dots \\ e^{2\pi i (M-1)f_0} & e^{2\pi i (M-1)f_1} & \dots & e^{2\pi i (M-1)f_{N-1}} \end{bmatrix}$$

with

$$f_n = \frac{1}{2\pi} \beta_n \delta = \sqrt{1 - n^2 \frac{\lambda^2}{4d^2}} \frac{\delta}{\lambda}$$

$\lambda$  : wavelength,  $d$  : waveguide thickness

**Invertibility of  $V^*V$  :** true if for all  $n' \neq n$ ,  $f_{n'} - f_n \notin \mathbb{Z}$

$\longrightarrow$  achieved for  $\delta \leq \lambda$

# Conditioning of $V^*V$

A classical question in **signal processing**...

**Condition number** :  $\kappa(V) = \sqrt{\sigma_{\max}(V^*V)/\sigma_{\min}(V^*V)}$

**Remark** : for  $n, n' = 0, \dots, N-1$ ,

$$(V^*V)_{nn'} = \sum_{m=0}^{M-1} e^{-2\pi i(f_n - f_{n'})m}$$
$$= \begin{cases} M & \text{if } f_n - f_{n'} \in \mathbb{Z} \\ \frac{1 - e^{-2\pi i(f_n - f_{n'})M}}{1 - e^{-2\pi i(f_n - f_{n'})}} & \text{if } f_n - f_{n'} \notin \mathbb{Z}. \end{cases}$$

**Uniform case** : if  $M/N \in \mathbb{N}$  and  $f_n = n/N$ ,  $n = 0, \dots, N-1$ , then

$$V^*V = M I_N$$

$\longrightarrow \kappa(V) = 1$  : optimal conditioning !



# Conditioning of $V^*V$

What about the general case ?

Wrap-around distance on  $[0, 1]$  :

$$d_w(f, g) = \inf_{q \in \mathbb{Z}} |f - g + q|$$

Example :  $d_w(0.1, 0.9) = 0.2$

**Theorem** (A. Moitra, 2015) : If the minimal separation

$$\Delta = \min_{n, n' = 0, \dots, N-1, n \neq n'} d_w(f_n, f_{n'})$$

is such that  $M > 1/\Delta + 1$ , then

$$\kappa(V) \leq \sqrt{\frac{M + 1/\Delta - 1}{M - 1/\Delta - 1}}$$

→ when  $M \rightarrow +\infty$ , optimal conditioning is recovered !

# The proof

**The Selberg's majorant and minorant :** for  $[a, b]$  and  $L > 0$ , there exist two functions  $S_+$  and  $S_-$  such that :

1.  $S_+$  and  $S_-$  are integrable
2.  $S_-(t) \leq \chi_{[a,b]}(t) \leq S_+(t)$  for all  $t \in \mathbb{R}$
3.  $\widehat{S}_\pm(x) = 0$  for  $|x| \geq L$

and

$$\int_{-\infty}^{+\infty} S_\pm(t) dt = b - a \pm \frac{1}{L}$$

Moreover

$$|S_\pm(t)| \leq \frac{C}{1+t^2}, \quad \forall t \in \mathbb{R}$$

→ The functions  $S_\pm$  can serve as test functions for the Dirac comb

# The proof (cont.)

We choose  $a = 0$ ,  $b = M - 1$  and  $L = \Delta$  in  $S_{\pm}$  :

Consider  $v(x) = \sum_{n=0}^{N-1} u_n e^{-2\pi i f_n x}$  for  $U = (u_0, u_1, \dots, u_{N-1}) \in \mathbb{C}^N$

$$\sum_{m=0}^{M-1} |v(m)|^2 = \sum_{n=0}^{N-1} \sum_{n'=0}^{N-1} u_n \bar{u}_{n'} \underbrace{\sum_{m=0}^{M-1} e^{-2\pi i (f_n - f_{n'}) m}}_{\delta_{nn'}}$$

$$= \sum_{n=0}^{N-1} \sum_{n'=0}^{N-1} (V^* V)_{nn'} u_n \bar{u}_{n'} = \|VU\|^2$$

Interpretation with **Dirac comb**  $D = \sum_{l \in \mathbb{Z}} \delta_l$  : since

$\chi_{[0, M-1]}(t) \leq S_+(t)$  for all  $t \in \mathbb{R}$

$$\sum_{m=0}^{M-1} |v(m)|^2 = \left\langle \chi_{[0, M-1]} D, |v|^2 \right\rangle \leq \left\langle D, S_+ |v|^2 \right\rangle$$

# The proof (cont.)

$$\begin{aligned}
 D = \sum_{l \in \mathbb{Z}} e^{-2\pi i l x} &\Rightarrow \sum_{m=0}^{M-1} |v(m)|^2 \leq \sum_{l \in \mathbb{Z}} \left\langle e^{-2\pi i l x}, S_+ |v|^2 \right\rangle \\
 &= \sum_{n=0}^{N-1} \sum_{n'=0}^{N-1} \sum_{l \in \mathbb{Z}} u_n \bar{u}_{n'} \underbrace{\int_{\mathbb{R}} e^{-2\pi i l x} S_+(x) e^{-2\pi i (f_n - f_{n'}) x} dx}_{\hat{S}_+(f_n - f_{n'} + l)}
 \end{aligned}$$

**Minimal separation :**  $\Delta = \inf_{n \neq n'} \inf_{l \in \mathbb{Z}} |f_n - f_{n'} + l|$  and the support of  $\hat{S}_+$  is contained in  $[-\Delta, \Delta]$

$$\sum_{m=0}^{M-1} |v(m)|^2 \leq \hat{S}_+(0) \sum_{n=0}^{N-1} |u_n|^2, \quad \hat{S}_+(0) = \int_{-\infty}^{+\infty} S_+(t) dt = M - 1 + 1/\Delta$$

**Final estimate :**  $\|VU\|^2 \leq (M - 1 + 1/\Delta) \|U\|^2$

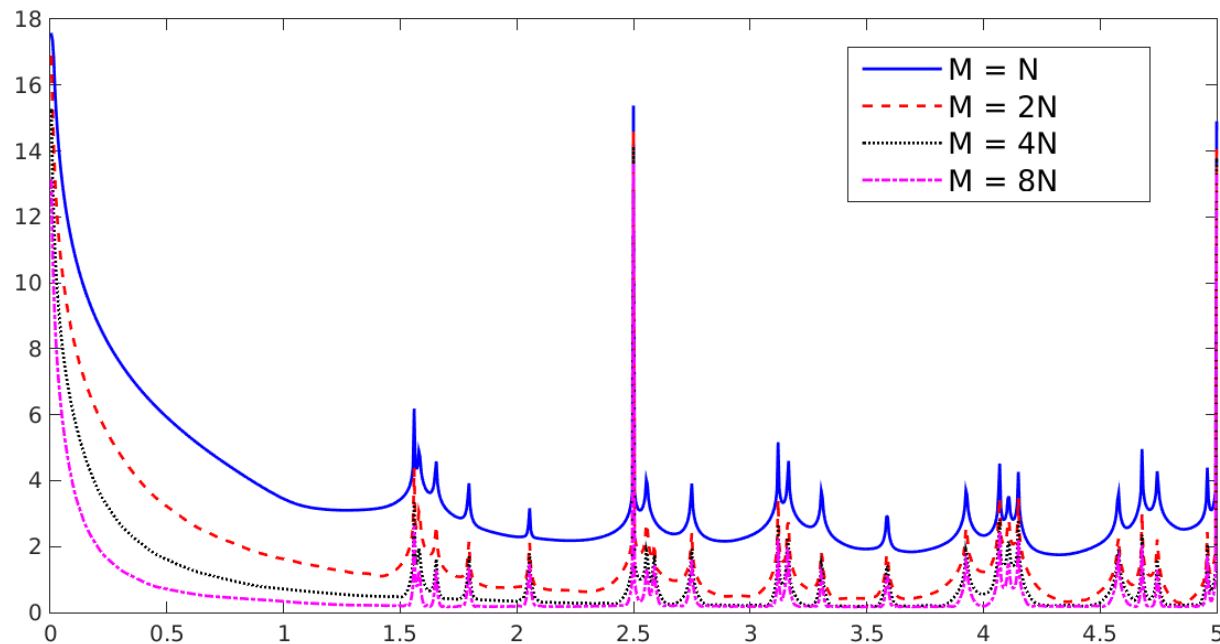
# Optimization of sources/receivers

**Conclusion :**  $M$  and  $\Delta$  have to be large !

In our particular case :

$$\Delta = \left(1 - \sqrt{1 - \frac{\lambda^2}{4d^2}}\right) \frac{\delta}{\lambda} \quad \longrightarrow \quad \text{Optimal } \delta \text{ is } \lambda$$

Log of the condition number  $\kappa(V)$  versus  $\delta/\lambda$  :



# Back to the time domain problem

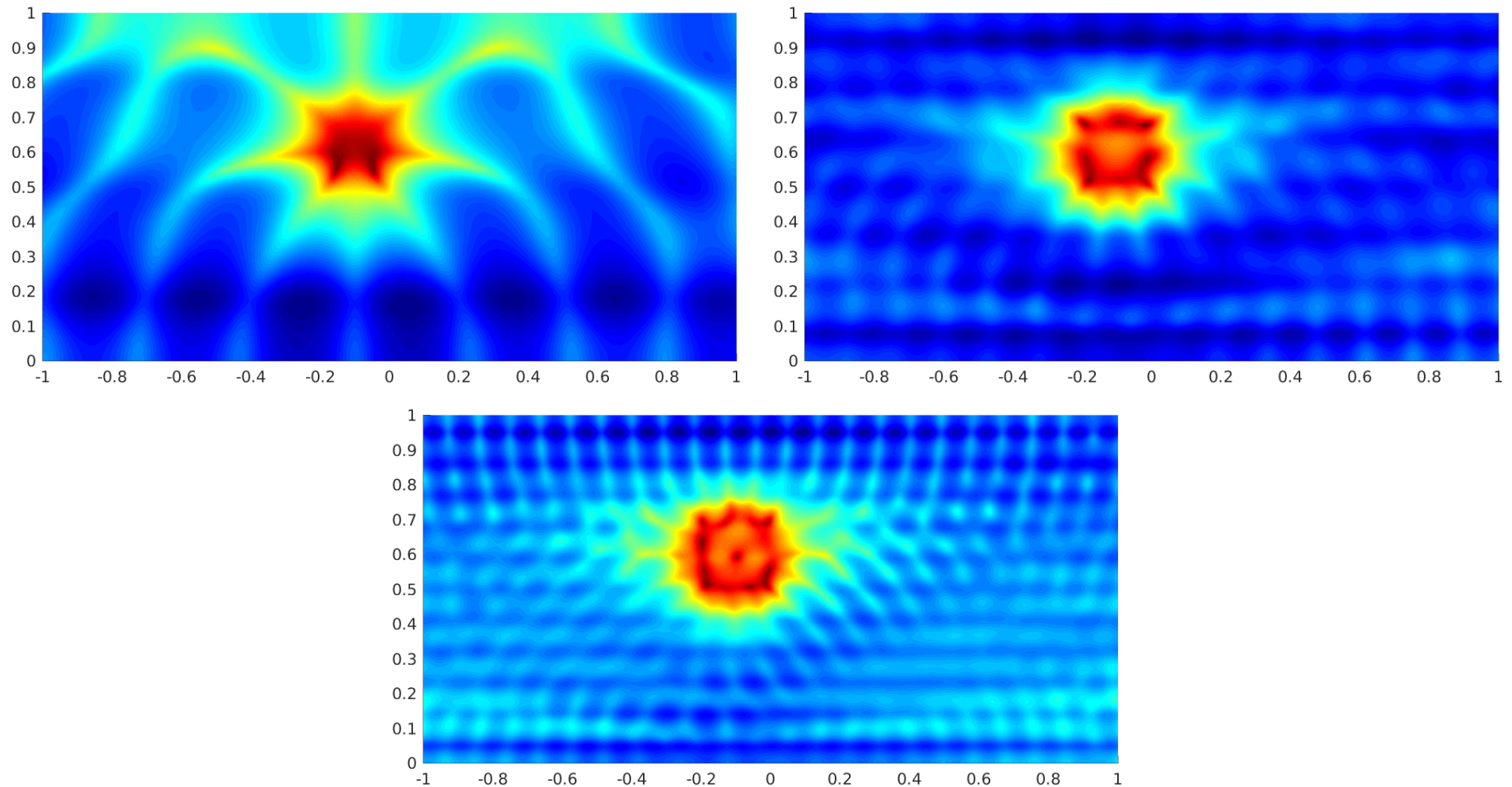
## Methodology:

- Take the Fourier transform of the data  
→ Measurement matrix  $M(\omega)$
- Invert the matrices  $E$  and  $R$  for each frequency  $\omega$   
→ LSM matrix  $U(\omega)$
- Solve the LSM modal formulation for each frequency  $\omega$   
→ Indicator function  $\psi(\omega)$
- Compute a global indicator function  $\Psi$  :

$$\Psi = \left( \int_{\omega_-}^{\omega_+} \frac{\max_{z \in G} |\psi(z, \omega)|^2}{|\psi(\cdot, \omega)|^2} d\omega \right)^{-1/2}$$

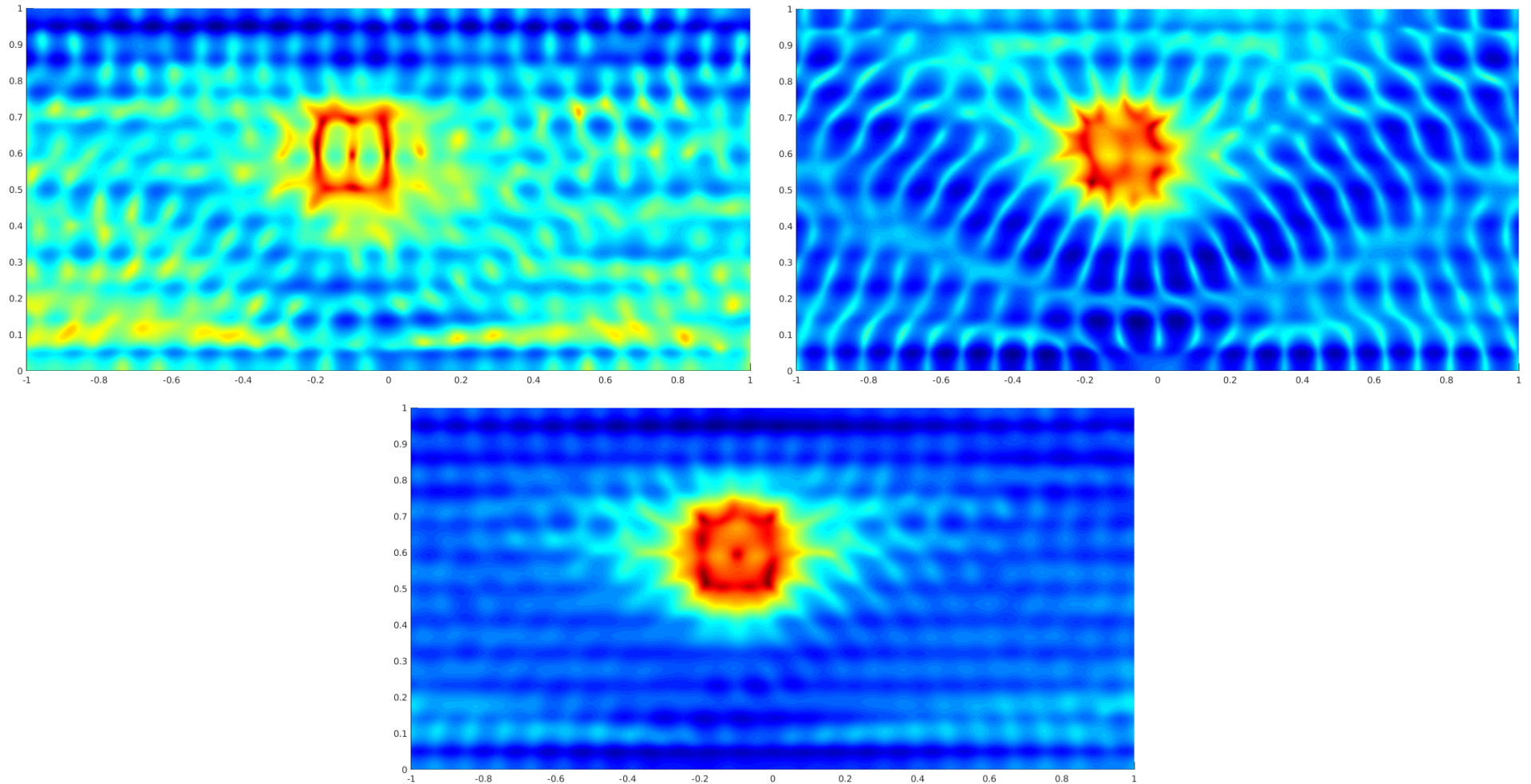
$G$  : sampling grid

# Some numerical results : time domain



Inversion using one single frequency corresponding to  $N = 4$ ,  
 $N = 8$  and  $N = 12$

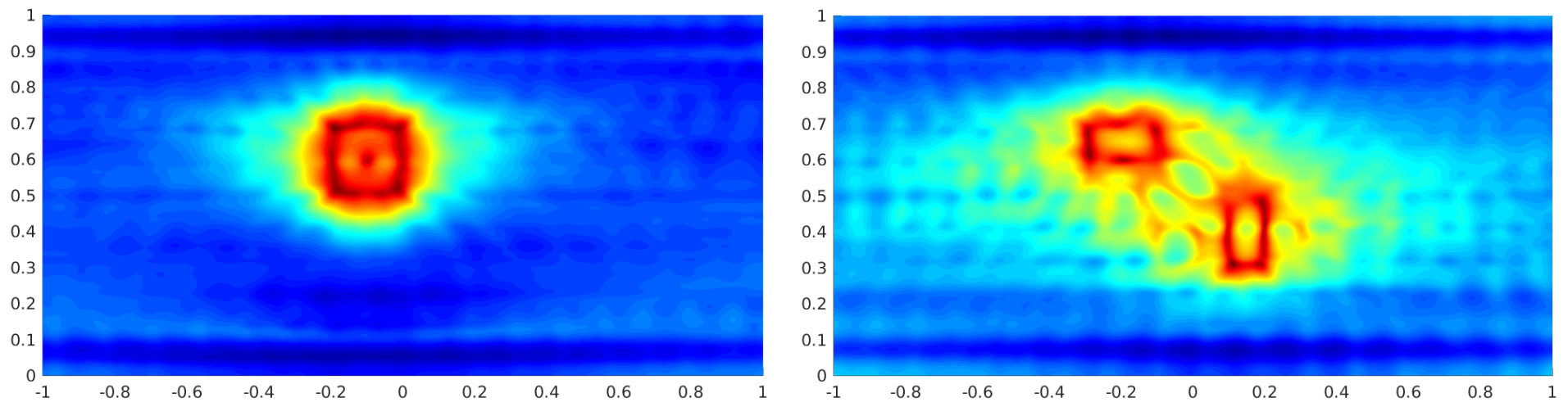
# Impact of the number of sources/receivers



Inversion using a single frequency corresponding to  $N = 12$  :  
 $M = N$ ,  $M = 2N$  and  $M = 3N$



# Some numerical results : time domain



Inversion using multiple frequencies

# Some perspectives

- Extension to elasticity (NDT by ultrasonics) : the modal approach is more difficult
- Extension to more realistic defects : corrosion, cracks
- Experimental validation with true data

Thank you for your attention !

# About the synthetic data

**Some difficulties** related to the time domain:

- From a numerical point of view, the forward computation is not that easy, many choices  $\longrightarrow$  Our choice : FE in space, FD in time, PMLs,...
- In a waveguide, the scattering field is slowly decaying with respect to time  $\longrightarrow$  the forward computation has to be long !

**Possible remedy** :  $\hat{\chi}(\omega)$  should avoid the cut-off frequencies  $\lambda_n$

